

Dual-mode resonator filter with improved feed-lines for dual-band applications

Shan Shan Gao^{1,2,3}, Jia-Lin Li^{4*}, Zhe Lin Zhu¹, Jia Li Xu¹, Yong Xin Zhao¹

An improved feedline configuration for dual-mode resonator filter is investigated in this paper. Based on the introduced topology, a dual-mode dual-band bandpass filter with center frequencies of 1.8 and 2.4 GHz is optimally designed, fabricated and tested. The introduced dual-band bandpass filter has simple structure and enables high selectivity to be realized due to two pairs of transmission zeros located near to the lower and upper passband, respectively. Both measured and simulated performances are presented with good consistency.

Key words: dual mode, dual band, bandpass, filter.

1 Introduction

The ever increasing demand for modern wireless communications necessitates efficient utilization of more and more frequency channels, which enables RF transceivers operating in multiple separated frequency band such that users can access various services with a single multimode handset. In such systems, dual or multi band filters play an important role. In [1], a dual-band filter was initially reported, where two single band bandpass filters (BPFs) are stacked together, one on the upper four layers and the other on the lower four layers with each filter having its own input and output ports. Based on slotlines, H. Zhang *et al* [2] etched slotlines on the edge sides of the substrate integrated cavities to achieve dual band, V. Nocella *et al* [3] designed a dual band filter, centered at 4.25 and 4.55 GHz respectively, by using TM dielectric loaded dual mode cavities. Further design for this kind of filters is based on the stepped impedance resonators (SIRs) [4-6]. One of the advantages for SIR filters is that the positions of the dual band can be tuned conveniently with the help of SIRs' inherent properties. Using magnetically coupled SIR, A. Lahmissi *et al* introduced this kind of topology [4]. Also, by using the common grounded via hole and the folded $\lambda/4$ SIRs, a high selectivity dual-band filter has been realized [5]. By applying defected split-ring resonators and irregular SIRs, a dual-band filter is reported in [6], where three transmission zeros are, respectively, located near to the two pass-band thus greatly improving the selectivity.

On the other hand, the dual mode microstrip resonators, as we known, are attractive since each of dual mode resonators can be used as a double tuned resonant

circuit, therefore, the number of resonators required for a given degree filter is reduced by half, resulting in a compact filter topology. Hence, P. Ma *et al* [7] introduced this kind of dual-mode dual-band filters with hairpin resonator. However, observing these dual-band filters, drawbacks are suffered. For example, the filter in [1] needs extra matching networks, whereas the in-band or out-of-band performance is not so good [3-7].

Thus, in this Letter, we investigate a dual-mode resonator filter designed for dual band applications. After simply analyzing the introduced basic topology, a dual mode dual band filter, centered at 1.8 and 2.4 GHz, respectively, is optimally designed, fabricated, and tested. Meanwhile, two pairs of transmission zeros, located near to the lower and upper passband, respectively, enable the introduced filter having high selectivity. The measured performances for the fabricated filter show good consistency with the simulated results.

2 Design of the dual mode resonator filter

As compared to the conventional dual-mode loop resonator filter, seen Fig. 1(a), the basic schematic diagram for the investigated filter topology is shown in Fig. 1(b), where a pair of feed-lines is symmetrically and spatially separated on the resonator. Thus in-line input and output (I/O) ports are realized this can increase the design flexibility. Though the I/O ports are not separated by 90° , two degenerate modes, here corresponding to TM_{100}^z TM_{010}^z fundamental mode applications (where z is perpendicular to the ground plane), can be also excited and coupled to each other due to the perturbation element

¹School of Electronic Information and Electrical Engineering, Chengdu University, Chengdu 610106, China, ²State Key Laboratory of Millimeter Waves, Nanjing 210096, China, ³Key Laboratory of Pattern Recognition and Intelligent Information Processing, Institutions of Higher Education of Sichuan Province, Chengdu University, Chengdu 610106, China, ⁴School of Physics, University of Electronic Science and Technology of China, Chengdu 610054, China, E mail: jialinli@uestc.edu.cn

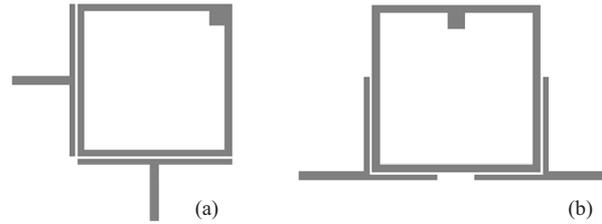


Fig. 1. Dual-mode loop resonator filter (a) – conventional topology (b) – introduced topology

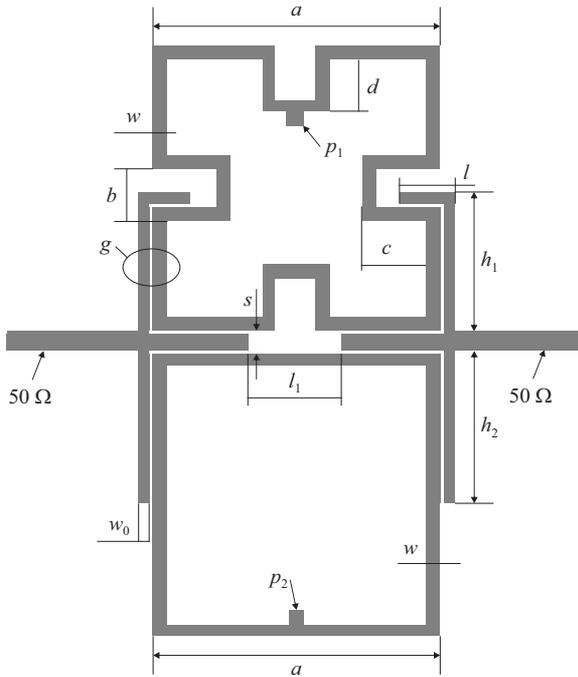


Fig. 2. Schematic diagram for the investigated dual-mode dual-band resonator filter passband performances; and (c) – upper passband performances.

within the resonator. The resonance for both cases can be established when the mean circumference L of the resonator is equal to an integral multiple of a guided wavelength, given by

$$L = n\lambda_g, \tag{1}$$

where n is the mode number and λ_g is the guided wavelength.

For dual band applications, we set two resonators with equal side length a , and then put together with a space s , shown in Fig. 2. A common feed line is shared between the resonators and the remains are coupled the upper and lower resonators, respectively. With this arrangement, two pairs of transmission zeros can be realized near to the lower and upper passband, which enhances the filter selectivity. The design procedure for this kind of filter can be summarized as follows: firstly, we design a square loop resonator that has a resonance frequency of 2.4 GHz, and then by meandering or fractal technique, we design another resonator 1.8 GHz. It is worth mentioning that the two resonators have the same side length a , and this is the key point to design this kind of dual-band resonator

filter. Finally, we place the two resonators together and introduce the proposed feed-line configurations. Thus, a dual-mode dual-band bandpass filter, as seen in Fig. 2, is realized.

Table 1. Dimension parameters for the proposed dual mode dual band filter

Parameters	(mm)	Parameters	(mm)
a	13.13	g	0.2
b	3.0	s	1.0
c	2.96	l_1	4.45
d	2.46	h_1	6.285
w	0.5	h_2	7.0
p_1	0.65	l	2.26
p_2	0.3	w_0	0.4

3 Simulated and measured results

The investigated dual-band filter has been optimally designed, fabricated, and tested. Its dimension parameters are given in Tab. 1. The simulation is accomplished using Ansoft Ensemble 8.0, a full-wave EM of moment (MoM). Measurements are carried out on an HP8510C network analyzer. Figure 3 shows the simulated and measured performances. The fabricated filter has minimum passband insertion losses of 3.35 and 3.68 dB at the center frequencies of 1.8 and 2.44 GHz, respectively. Two pairs of transmission zeros located at 1.725, 1.97, 2.36 and 2.625 GHz, respectively, have been realized with the attenuation levels over 45 dB. Compared to the results given in [3, 4] and [7], these transmission zeros have greatly improved the filter selectivity, especially, for the upper passband. Though the slight shift in frequency between simulations and measurements due to the fabrication uncertainties, good consistency between them can be observed.

4 Conclusions

Based on the improved in-line feed-line configuration, a dual mode dual-band filter has been investigated in this letter. Centered at 1.8 and 2.4 GHz, respectively, a demonstrator filter has been optimally designed and examined with good consistency between simulations and measurements. Its enhanced selectivity and

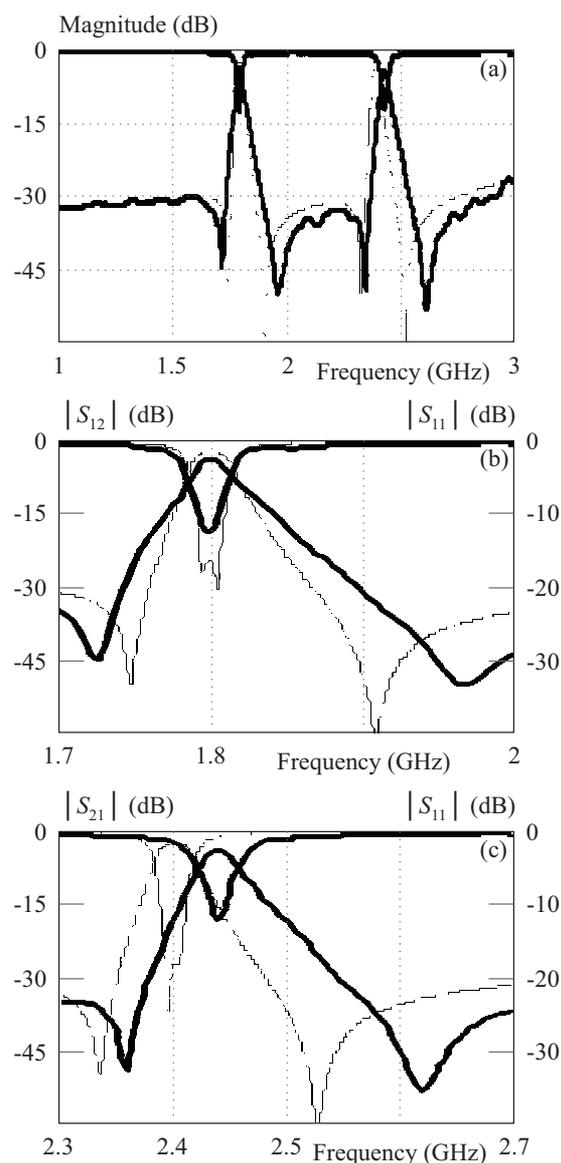


Fig. 3. Measured and simulated performances (measured simulated): (a) – broadband responses; (b) – lower

simple topology are attractive in modern wireless communications. Acknowledgment: This work was supported in part by the National Natural Science Foundation of China (NSFC) (61601063), in part by Open Foundation of State Key Laboratory of Millimeter Waves, the Southeast University (K202121), in part by Open Fund of Sichuan Provincial Key Laboratory of Pattern Recognition and Intelligent Information Processing of Chengdu University (MSSB-2020-08).

REFERENCES

[1] H. Miyake, S. Kitazawa, T. Ishizaki, T. Yamada, and Y. Nagatomi, “A miniaturized monolithic dual band filter using ce-

ramic lamination technique for dual mode portable telephones”, *1997 IEEE MTT-S Int. microw. Symp. Dig.*, vol. 2, pp. 789-792, 1997.

- [2] H. Zhang, W. Kang, and W. Wu, “Miniaturized dual-band SIW filters using E-shaped slotlines with controllable center frequencies”, *IEEE Microw. Wirel. Comp. Lett.*, vol. 28, no. 4, pp. 311-313, 2018.
- [3] V. Nocella, L. Pelliccia, C. Tomassoni, and R. Sorrentino, “Miniaturized dual-band waveguide filter using TM dielectric-loaded dual-mode cavities”, *IEEE Microw. Wirel. Comp. Lett.*, vol. 26, no. 5, pp. 310-312, 2016.
- [4] A. Lahmissi and M. Challal, “A novel microstrip dual-band bandpass filter with wide and deep stopband using modified SIRs”, *2017 5th International Conference on Electrical Engineering-Boumerdes (ICEE-B)*, pp. 1-4, 2017.
- [5] D. Wang, Z. He, and H. Yang, “Compact dual-band filter based on quarter-wavelength SIRs with grounded via coupling”, *2019 IEEE Asia-Pacific Microwave Conference (APMC)*, pp. 774-776, 2019.
- [6] X. Luo, X. Cheng, J. Han, L. Zhang, F. Chen, Y. Guo, X. Xia, and X. Deng, “Compact dual-band bandpass filter using defected SRR and irregular SIR”, *Electron. Lett.*, vol. 55, no. 8, pp. 463-465, 2019.
- [7] P. Ma, B. Wei, J. Hong, X. Guo, B. Cao, and L. Jiang, “Coupling matrix compression technique for high-isolation dual-mode dual-band filters”, *IEEE Trans. Microw. Theory Tech.*, vol. 66, no. 6, June 2006, pp. 2814-2821, 2018.

Received 24 October 2020

Shan Shan Gao was born in Sichuan Province, China. She received the PhD degree in radio physics from the University of Electronic Science and Technology of China (UESTC), Chengdu, in 2012. From June 2011 to July 2013, she was a Research Assistant at the Department of Electrical & Electronic Engineering, the University of Hong Kong. Currently, she is an Associate Professor at the School of Electronic Information and Electrical Engineering, Chengdu University.

Jia-Lin Li was born in Sichuan Province, China. He received the MS degree from the University of Electronic Science and Technology of China (UESTC), Chengdu, China, in 2004, and the PhD degree from the City University of Hong Kong, Hong Kong, in 2009, both in electronic engineering. From September 2005 to August 2006, he was a Research Associate with the Wireless Communication Research Center, City University of Hong Kong, Hong Kong. Since September 2009, he has been with the School of Physics, UESTC, where he is currently a Professor. His research interests include microwave/millimeter-wave antenna and arrays, circuits and systems, interactions between microwave and complex medium, and so on.

Zhe Lin Zhu was born in Sichuan Province, China. He is an undergraduate student at the School of Electronic Information and Electrical Engineering, Chengdu University.

Jia Li Xu was born in Sichuan Province, China. She is currently an Associate Professor at the School of Electronic Information and Electrical Engineering, Chengdu University.

Yong Xin Zhao was born in Sichuan Province, China. She is an experimentalist at the School of Electronic Information and Electrical Engineering, Chengdu University.